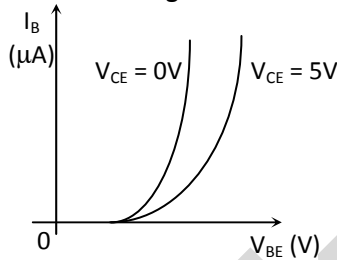


**Q.1(a) Draw input & output characteristics of BJT. State significance of DC load line. [5]**

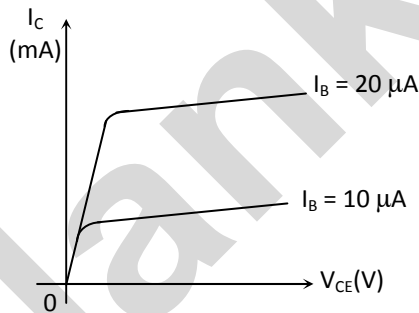
Ans.: (a) Input characteristics

$$V_{BE} = f(I_B, V_{CE})$$



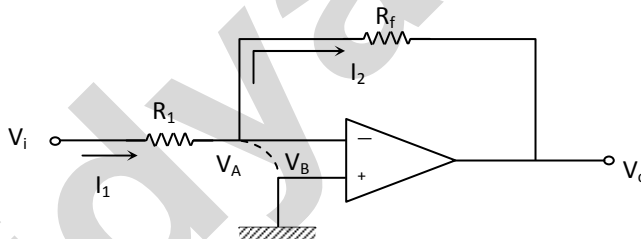
(b) Output characteristics

$$I_C = f(I_B, V_{CE})$$



**Q.1(b) Explain the concept of virtual ground in operational amplifiers. [5]**

Ans.:



Consider an inverting buffer amplifier as shown. Let  $V_a$  the potential at the inverting terminal of OP-AMP, assume that the input current of OP-AMP = 0, applying KCL at the inverting terminal node.

$$I_1 = I_2$$

$$\frac{V_i - V_a}{R_i} = \frac{V_a - V_o}{R_f}$$

$$V_i - V_a = V_a - V_o$$

The circuit is inverting buffer  $A = -1$

So  $V_o = AV_i = -V_i$

$$V_i - V_a = V_a - (-V_i)$$

$$2V_a = V_i - V_i = 0 \quad V_a = 0$$

Very few OP-AMPS for eg. 301, 748 & 777 have the OFFSET. Voltage null capability. Hence for most of the OP-AMPS we have to design an OFF-SET voltage compensation network in order to reduce the output offset voltage to zero.

**Q.1(c) Compare class A and class C amplifier.**

[5]

**Ans.:** In this **class-A**, amplifier operates in active region at all the times( $2\pi$ ).

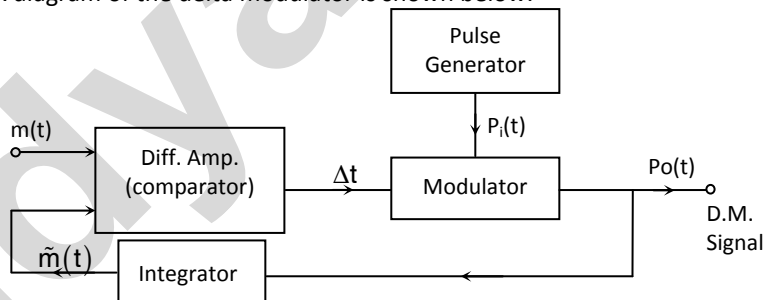
- Stage efficiency =  $P_{oAC} / P_{DC} \times 100\%$   
Conduction angle  $2\pi$   
Conversion efficiency = 50%
- Load Power,  $P_L = (V_L)^2 / R_L = (V_{PP})^2 / 8 \cdot R_L$
- In this **class-C**, amplifier operates for less than half of the input cycle (less than  $\pi$ ).
- It is basically a tuned amplifier.
- AC load power for class-C amplifier is,  $P_L = (V_{PP})^2 / 8 \cdot R_L$   
Where,  $V_{PP}$  is peak to peak load voltage and  $R_L$  is load resistance  
Conduction angle less than  $\pi$   
Conversion efficiency  $\approx 90\%$

**Q.1(d) Explain adaptive delta modulation.**

[5]

**Ans.:** Delta modulation is also a digital modulation system. In delta modulation just one bit is sent per sample, to indicate whether the present sample is larger or smaller than the previous sample. Thus in delta modulation, the modulated signal carries information not about the signal samples but about the difference between successive samples. If the difference is +ve or -ve a +ve or -ve pulse respectively is generated in the modulated signal. Thus D.M. carries the information about the derivative of  $m(t)$  and hence the name is delta modulation. Thus integration of delta modulated signal  $m(t)$  will be an approximation of  $m(t)$ .

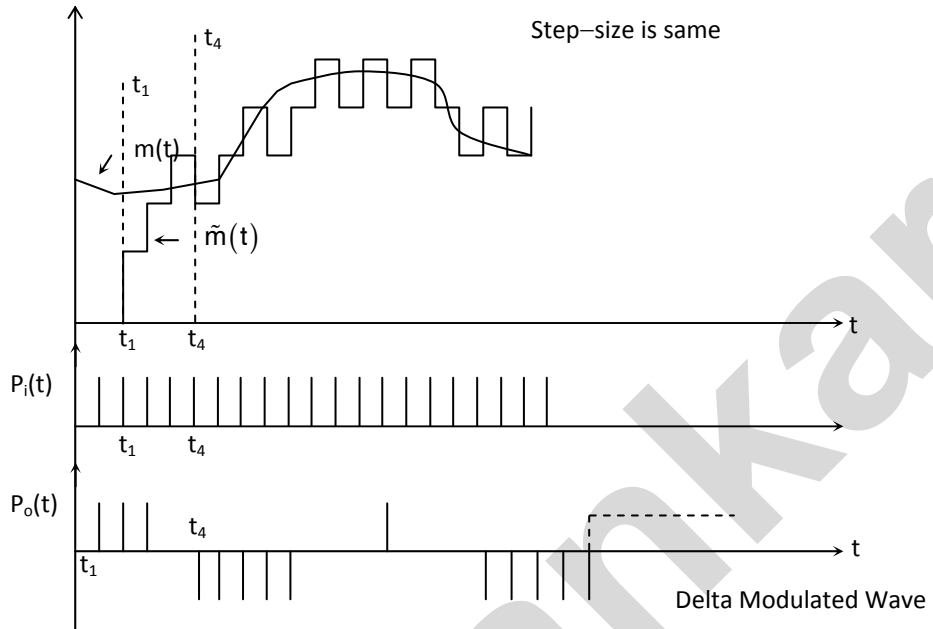
The block diagram of the delta modulator is shown below.



The pulse generator produces a pulse train  $P_i(t)$  of positive pulses. The modulator receives  $P_i(t)$  and  $\Delta t$ , the output of differential amp. The modulator output  $P_o(t)$  is the product of input pulse train  $P_i(t)$  and  $+1$  or  $-1$  depending upon the polarity of  $\Delta t$ .  $P_o(t)$  is positive pulse if  $\Delta t$  is +ve and it is -ve pulse if  $\Delta t$  is -ve. The magnitude of  $\Delta t$  has no role in deciding  $p_o(t)$ . The output of modulator is applied to an integrator whose output is  $\tilde{m}(t)$ .

The input signal  $m(t)$  and the integrator output  $\tilde{m}(t)$  is compared in difference amp. Whose output is  $\Delta(t) = m(t) - \tilde{m}(t)$ .

Following figure explains the operation of delta modulator.



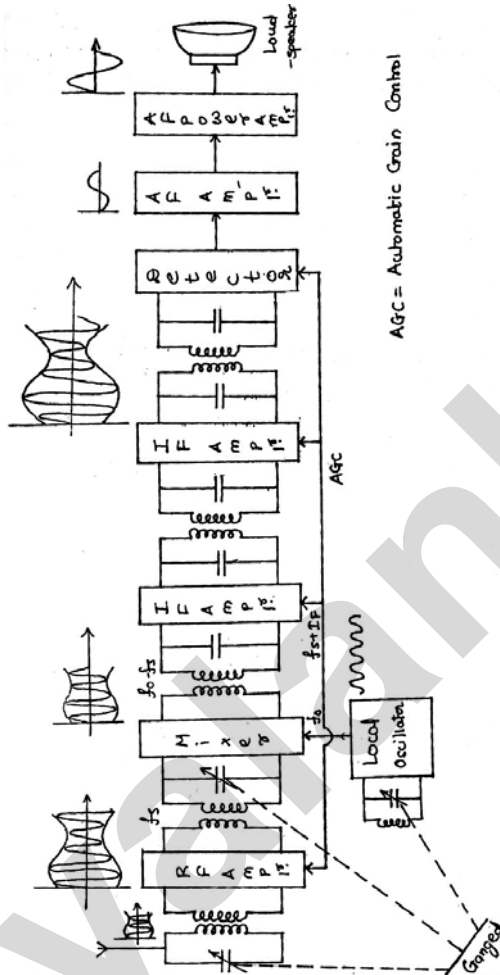
**Q.2(a) Explain Superheterodyne receiver with suitable diagram. [10]**

**Ans.:** The features of heterodyne receiver is that all incoming radio frequency signals called intermediate frequency by using heterodyne process.

In this receiver, the received signal frequency is mixed with the local oscillator frequency  $f_0$ . The local oscillator is an LC oscillator which produce sinusoidal oscillations of frequency  $f_0$ . The frequency of local oscillator depends on the values of L & C of associated tuned circuit. The signal frequency  $f_s$  and signal  $f_0$  gives to the mixer circuit.

Mixer is a non-linear circuit. It will produce the output which contain different frequency component such as  $f_s$ ,  $f_0$ ,  $f_0 \pm f_s$ , harmonics of input frequencies and the IC components. The inputs tank circuit of IF amplifier is tuned to  $f_0 - f_s$  which is always 455 KHz and is known as Intermediate frequency  $I_f$ . To get  $I_f$  always = 455 KHz, the local oscillator should be tuned such that whatever may be input signal frequency  $f_s$ , local oscillator frequency  $f_0$  should be always 455 KHz, more than  $f_s$ . AM wave with 455 KHz is amplified to the desired level by 2 or 3  $I_f$  amplifier and then it is fed to the detector which detects a taken out the AF information from the AM wave. This audio frequency signal is then amplified by voltage and power amplifier so that it can drive the loudspeaker.

**Block diagram of Superheterodyne Receiver**



AGC = Automatic Gain Control

**Q.2(b)** The emitter bias configuration as shown in following figure has the specifications:

$$I_{CQ} = \frac{1}{2} I_{Csat}, I_{Csat} = 8 \text{ mA}, V_C = 18\text{V} \text{ and } \beta = 110$$

**Determine  $R_C$ ,  $R_E$  and  $R_B$ .**

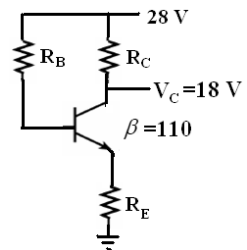
**Ans.:** Given :  $I_{CQ} = 4 \text{ mA}$ ,  $\beta = 110$ ,  $V_C = 18\text{V}$

Apply KVL to collector,

$$28 - I_C R_C - 18 = 0$$

$$R_C = \frac{28 - 18}{I_C} = \frac{10}{4 \text{ mA}} = 2.5 \text{ K}\Omega$$

$$I_{Csat} = \frac{V_{CC}}{R_C + R_E}$$



[10]

$$\Rightarrow 8 \text{ mA} = \frac{28}{2.5\text{K}\Omega + R_E}$$

$$R_E = 1 \text{ k}\Omega$$

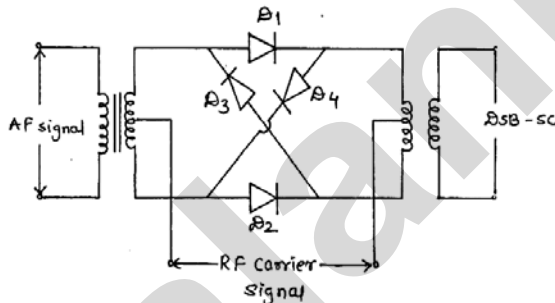
$$I_B = \frac{I_C}{\beta} = 36.3 \mu\text{A}$$

$$I_B = \frac{V_{CC} - V_{BE}}{R_B + (1 + \beta)R_E} = \frac{28 - 0.7}{R_B + (1 + 110)}$$

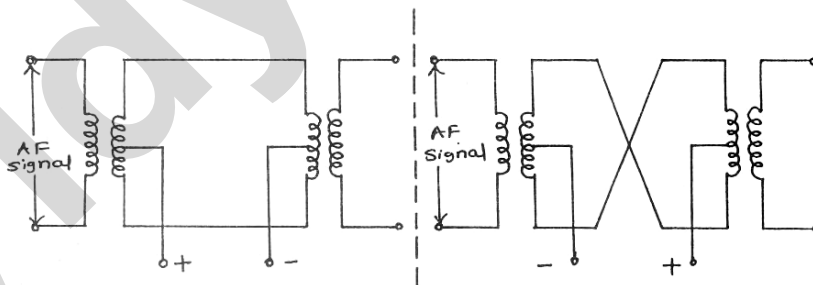
$$\therefore R_B = 639.82 \text{ k}\Omega$$

**Q.3(a) Explain generation of DSBSC using Ring Modulator. [10]**

**Ans.:** The ring modulator is also known as Lattice or double balanced modulator. In this circuit, four diodes are connected in same direction and they will form a ring. Hence the name Ring Modulator.



During +ve half cycle of carrier signal diode  $D_1$  and  $D_2$  are forward biased and acts as short circuit and diode  $D_3$  and  $D_4$  are reverse biased and act as open circuit as shown in figure A. Therefore, the modulating signal multiplies with the carrier signal, in the primary of output transformer.

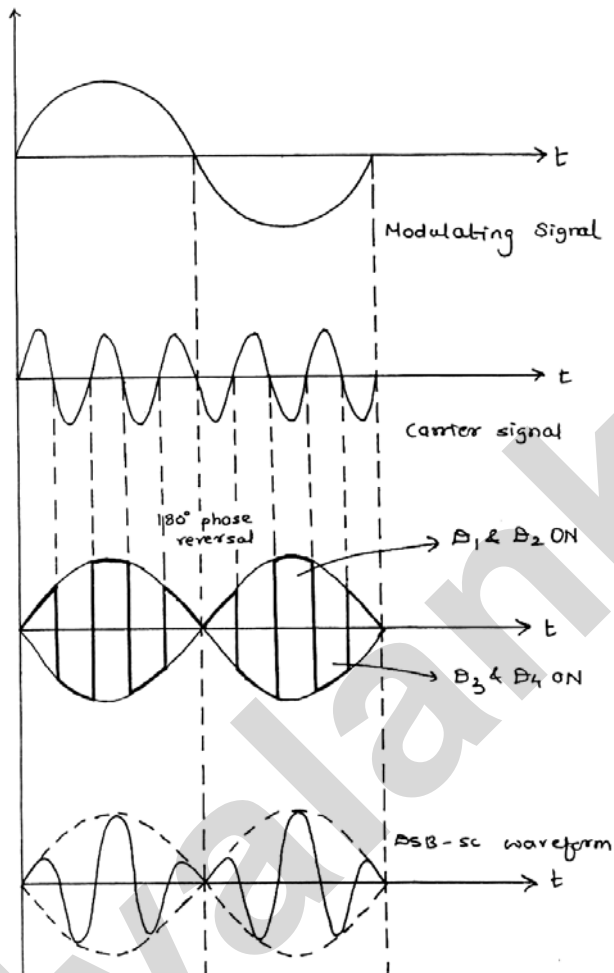


During +ve half cycle of carrier.

**Fig. (a)**

During -ve half cycle of carrier.

**Fig. (b)**



During  $-ve$  half cycle of carrier signal. Diode  $D_3$  and  $D_4$  are F.B. and act as short circuit and diode  $D_1$  and  $D_2$  are R.B. and act as open circuit as shown in figure B. Therefore, the modulating signal  $m(t)$  is reversed i.e.  $-m(t)$  is multiplied by the carrier in the primary of output transformer. Thus the resultant current in primary is the subtraction of current through primary during  $+ve$  half cycle and  $-ve$  half cycle of RF signal. Therefore, at secondary of output transformer. the carrier is suppressed and we get only DSB-SC.

**Q.3(b) With suitable waveforms explain how Op-amp can be used as integrator. [10]**

**Ans.:** A circuit in which the output voltage waveform is the integral of the input voltage waveform is the integrator or the integration amplifier. Such a circuit is obtained by using a basic inverting amplifier configuration if the feedback resistor  $R_F$  is replaced by a capacitor  $C_F$ .

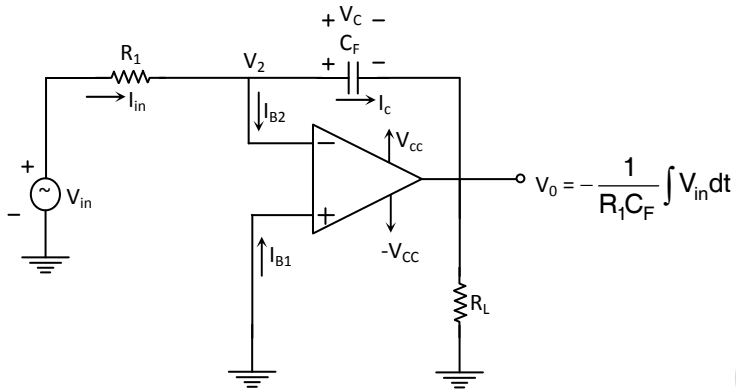


Fig. (a) : Basic Integrator

**Expression for output of basic integrator Fig.(a)**

The expression for output  $V_0$  can be obtained by writing KCL equation at node  $V_2$ .

$$I_{in} = I_{B2} + I_c$$

Since  $I_{B2} = 0$

$$I_{in} = I_c$$

$$\frac{V_{in} - V_2}{R_1} = C_F \frac{dV_c}{dt} \quad (\because V_c = V_2 - V_0)$$

$$\frac{V_{in} - V_2}{R_1} = C_F \frac{d(V_2 - V_0)}{dt}$$

However,  $V_1 = V_2 = 0$ , because A is very large.

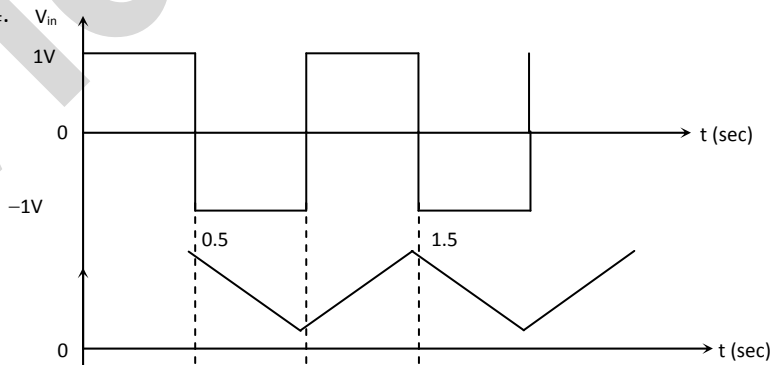
$$\frac{V_{in}}{R_1} = C_F \frac{dV_0(-)}{dt}$$

Integrate both sides with respect to time t.

$$\int \frac{V_{in}}{R_1} dt = C_F \int \frac{d}{dt} (-V_0)$$

$$V_0 = -\frac{1}{R_1 C_F} \int V_{in} dt \quad \dots (A)$$

The equation (A) shows that output voltage is directly proportional to time constant  $R_1 C_F$ .



- Q.4(a) For an AM DSBFC envelope with  $V_{\max} = 20V$  and  $V_{\min} = 4V$ ; determine :** [10]
- (i) Peak amplitude of USF and LSF
  - (ii) Peak amplitude of carrier
  - (iii) Peak change in the amplitude of envelope
  - (iv) Modulation coefficient
  - (v) Draw the AM Envelope

**Ans.:** Given :  $V_{\max} = 20V$  and  $V_{\min} = 4V$

- (ii) Peak amplitude of carrier

$$V_c = \frac{V_{\max} + V_{\min}}{2} = \frac{20 + 4}{2} = \frac{24}{2} = 12 V$$

$$V_m = \frac{V_{\max} - V_{\min}}{2} = \frac{20 - 4}{2} = \frac{16}{2} = 8 V$$

- (iii) Peak change in the amplitude of envelope  $\pm 4V$

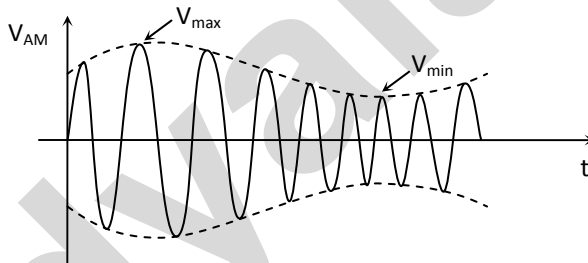
- (iv) Modulation coefficient

$$m = \frac{V_m}{V_c} = \frac{8}{12} = 0.666$$

- (i) Peak amplitude of USF and LSF

$$A_{USB} = A_{LSB} = \frac{mV_c}{2} = 3.96 V$$

- (v) Draw the AM Envelope



- Q.4(b) Draw the PAM, PPM and PWM waveforms and compare them.** [5]

**Ans.:**

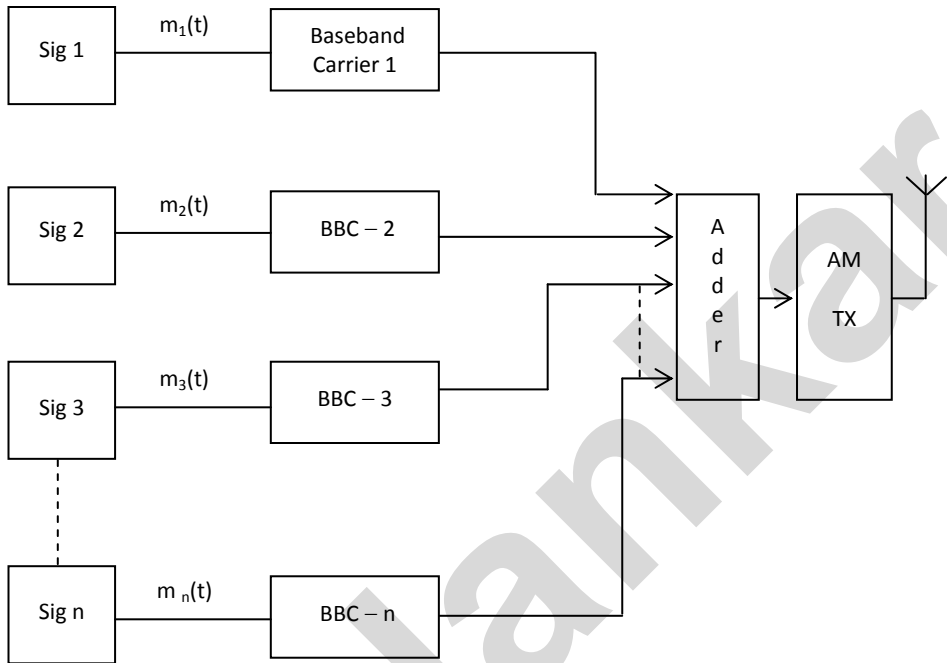
	Parameter	PAM	PWM	PPM
(i)	Information contained in	Amplitude Variation	Width Variation	Position Variation
(ii)	Bandwidth Requirement	low	high	high
(iii)	Noise immunity	low	high	high
(iv)	Transmitted Power	Varies with amplitude & pulses	Varies with variation in width	Remains constant
(v)	Need to transmit synchronization pulses	not needed	not needed	necessary
(vi)	Complexity of generation & detection	Complex	Easy	Complex



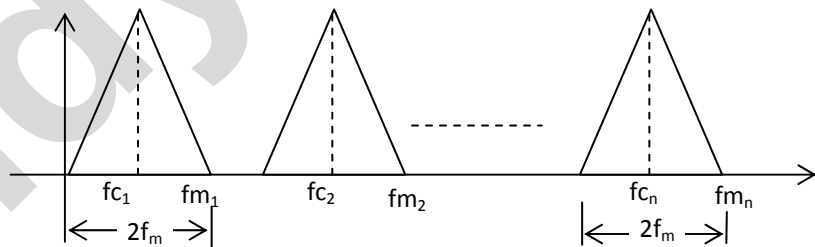
**Q.4(c) Explain principle of FDM.**

**[5]**

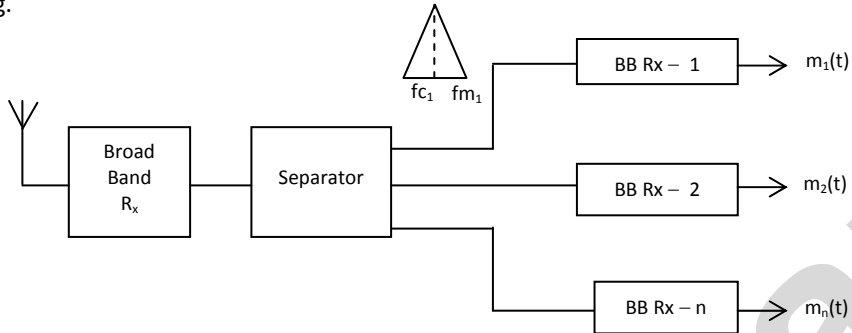
**Ans.:** FDM consists of simultaneous transmission of message of different channels by shifting them in frequency domain. The block diagram of FDM Tx & Rx is as shown below.



In above diagram, all the base band carrier blocks are the amplitude modulators. Here, all the signals have the same frequency spectrum. Therefore, all these signals are upshifted in different frequency slot using amplitude modulator. For every BBC Block, the carrier frequency is different. Hence the output of BBC - 1 will be  $f_{c1} \pm f_{m1}$ . The output of BBC - 2 will be  $f_{c2} \pm f_{m2}$ . If all the modulating signals have the same B.W. then output of Adder will have the following spectrum.



This FDM signal is transmitted by the Tx. The block diagram of FDM Rx is as shown in fig.



At the receiving station, the FDM signal is received by the broad band Rx. The separator circuit uses approximate mixers & filters, tuned at proper frequencies. The separator has n outputs & each o/p is the modulated signal centered around  $fc_1, fc_2 \dots fc_n$  respectively. These outputs are given to  $BBR_x$ . These are nothing but AM  $R_x$ s. Hence the output of each block is corresponding modulating signal.

**Q.5(a) Explain and give ideal values of following parameters of an Op-Amp : (i) CMRR, [10] (ii) Slew rate, (iii) Offset voltage, (iv) Input Resistance, (v) Output Impedance**

**Ans.: (i) CMMR :**

It can be defined as the ratio of the differential voltage gain  $A_d$  to the common mode voltage gain  $A_{cm}$  is,

$$CMRR = \frac{A_d}{A_{cm}}$$

$A_d$  is same as large signal voltage gain A. the common mode voltage gain is determined from the circuit of figure (a) using the equation,

$$A_{cm} = \frac{V_{ocm}}{V_{cm}}$$

where  $V_{ocm}$  = common mode output voltage

$V_{cm}$  = common mode input voltage

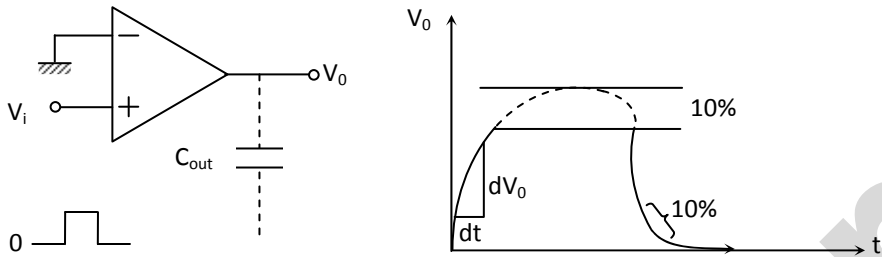
Generally, the  $A_{cm}$  is very small and  $A_d = A$  is very large, hence CMRR is very large and being a large value it is expressed in dB.

For 741C CMRR with (source resistance)  $R_s \leq 10k\Omega = 90dB$

The higher the value of CMRR, better is the matching between two input terminals and smaller is the output common mode voltage.

The 714C precision op-amp has CMRR = 120dB hence, it has better ability to reject common mode voltages such as electrical noise and preferred in noisy environment.

(ii) Slew rate:



When the OP-AMP is operated at high frequencies and when it is excited by a large signal voltage, due to saturation of the internal stages of OP-AMP, there is an effective capacitance  $C_{out}$  appearing across output terminals of OP-AMP ( $C_{out}$  is due to lead, wiring, stray and internal capacitance of OP-AMP).

Since voltage across capacitor cannot change instantaneously, it takes same time for output voltage to respond to input. Slow rate is defined as the maximum rate of change of output voltage w.r.t. to time.

$$S = \left. \frac{dV_o}{dt} \right|_{max} ; \text{Unit is V}/\mu \text{ sec}$$

For ideal OP-AMP, slow rate =  $\infty$ . For  $\mu A741$  slow rate =  $0.5 \text{ V}/\mu \text{ sec}$ .

Let  $V_o = V_m \sin(\omega_m t)$

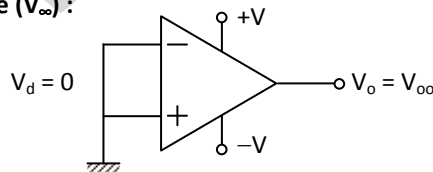
$$\begin{aligned} \frac{dV_o}{dt} &= V_m \omega_m \cos(\omega_m t) \\ &= 2\pi f_m V_m \cos(\omega_m t) = 2\pi f_m V_m \cos \theta \end{aligned}$$

$$\begin{aligned} S &= \left. \frac{dV_o}{dt} \right|_{max} = 2\pi f_{max} V_{max} 1 \\ &= 2\pi f_{max} V_{max} 1 \text{ V}/\mu \text{ sec} \end{aligned}$$

where,  $V_{max}$  = maximum output voltage OP-AMP can provide without distortion

$f_{max}$  = The maximum frequency OP-AMP can amplify without distortion

(iii) O/P Offset Voltage ( $V_{oo}$ ):



$V_{oo}$  is defined as the output voltage measured when the differential input voltage  $V_d$  is set zero. For ideal OP-AMP,  $V_{oo} = 0$ . Practically, it can be as high as supply voltage  $\pm V$ .

Due to the mismatch but the transistors used at the i/p stage of OP-AMP, there is a small differential o/p obtained from the first stage which is further amplified by the succeeding stages generating an output offset voltage  $V_{oo}$ .

If  $V_{oo} = 0$ , the OP-AMP is said to be balanced or nulled.

**(iv) Input resistance ( $R_i$ ) :**

It is defined as resistance which can be measured at either the inverting or non-inverting terminal with the other terminal connected to ground.

It is few mega  $\Omega$  for transistor input stage and extremely high (thousand of  $G\Omega$ ) for op-amp having FET based input stages.  $-R_i = 2M\Omega$ .

**(v) Output Resistance ( $R_o$ ) :**

Output resistance  $R_o$  is the equivalent resistance that can be measured between the output terminal of the op-amp and the ground. It is  $75\Omega$  for 741C op-amp.

**Q.5(b) Define and explain in brief Amount of information, average information, [10] information rate and Channel capacity of a communication system.**

**Ans.:** The more the probability of an event the less is the amount of information associated with it and vice versa.

Thus we can define "Information as reciprocal of probability"

$$\text{i.e. } I_k = \log_2 \left( \frac{1}{P_k} \right) \quad \dots (1)$$

$I_k \Rightarrow$  is the information of event 'k' with probability  $P_k$

The unit of information is 'bit' when the base of log is '2' for base 'e' the unit is 'hat' for base 10. Unit is decit or Harley.

"Actually information is dimensionless quantity"

**Average Information :** The "Entropy" is defined as average information per message. It is denoted by 'H' and its unit is bits/message.

It is given by

$$H = - \sum_{k=1}^M P_k \log_2 P_k \quad \dots (1)$$

This expression indicates that the entropy of a source is depends only on the probabilities.

**Information Rate :**

If a message source generates message is at rate of 'r' messages per second. Then the rate of information " 'R' is defined as average number of bits of information per second". [Average information per second]

We know 'H' is the average number of bits of information per message

Hence  $R = r.H$  bits/sec.

If the source of message generates 'r' number of messages per second then. Information rate is given by

$$R = r \times H \quad r \Rightarrow \text{No. of messages / sec}$$

$$H \Rightarrow \text{Message information / message}$$

$$R = \left[ r = \frac{\text{Message}}{\text{Sec}} \right] \times \left[ H = \frac{\text{Average Information}}{\text{message}} \right]$$

$$R = \text{Av. Information / Sec} \Rightarrow \text{bits / sec.}$$

**Channel Capacity**

The channel capacity is defined as the maximum possible bit rate a channel can support without introducing any error. i.e. channel capacity is the maximum of mutual information.

$$\therefore C = \max I(x; y) \text{ bits/sec}$$

The channel capacity is a function of only transition probabilities  $P(y_j/x_i)$  which define the channel.

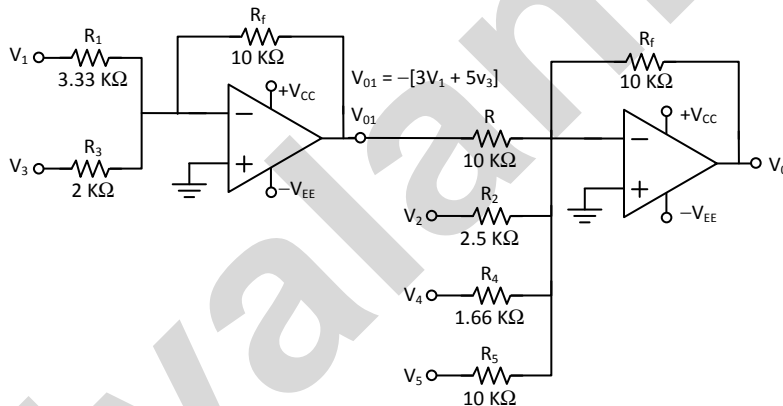
Channel capacity is defined as maximum average mutual information in any single use of the channel (i.e. signaling interval)

The transmission efficiency or channel efficiency is defined as

$$\eta = \frac{\text{Actual transinformation}}{\text{Maximum transinformation}}$$

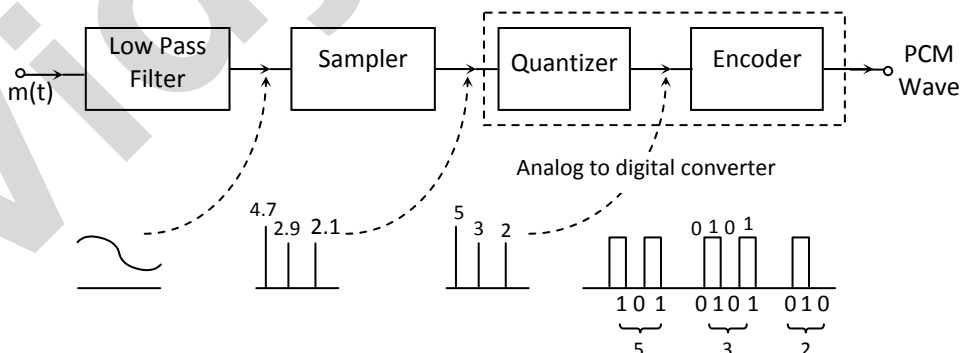
**Q.6(a) Implement using IC 741,  $V_o = 3V_1 - 4V_2 + 5V_3 - 6V_4 - V_5$**  [5]

**Ans.:**  $V_o = 3V_1 - 4V_2 + 5V_3 - 6V_4 - V_5$   
 $= (3V_1 + 5V_3) - (4V_2 + 6V_4 + V_5)$



**Q.6(b) Write a note on Pulse Code Modulation with waveforms.** [5]

**Ans.:** Following fig. shows the block diagram of the P.C.M. system.



**Sampler :** The incoming modulating signal is sampled with a train of narrow rectangular pulses. The sampling rate must be greater than twice the highest modulating frequency  $> 2 f_{m(max)}$ .

The low pass filter is used to avoid frequencies greater than  $f_{m(max)}$ .  
 The sampled pulses are then quantized in the quantizer [Explain quantization].  
 The encoder encodes these quantized pulses into bits [Binary Code] as shown in fig.

**P.C.M. B. W.**

B.W. =  $[n(N + 1) + 1] 2f_m$   
 where  $f_m$  = modulating frequency  
 $n$  = no. of channels or signals to be transmitted  
 $N$  = no. of bits per samples

**Q.6(c) Obtain Expression for an FM Wave. [10]**

**Ans.:** The instantaneous frequency of FM wave is given by

$$\omega t = \omega_c + \delta \quad \dots(1)$$

where  $\delta$  is variation above or below  $\omega_c$  depending on the modulating signal

i.e.  $\delta \propto V_m \cos \omega_m t$   
 $\delta = K V_m \cos \omega_m t \quad \dots(2)$

where  $K$  is frequency deviation sensitivity and is defined as ratio of maximum frequency deviation to the amplitude of modulating signal and is measure in Hertz per volt i.e. Hz/Volts

$$\delta_{max} = K V_m \quad \dots(3)$$

The instantaneous frequency of FM wave is given by

$$\omega_i = \omega_c + \delta \quad \dots(4)$$

Let FM wave be defined by

$$v_{fm}(t) = V_c \sin \theta \quad \dots(5)$$

where  $\theta$  is instantaneous phase and is obtained using instantaneous frequency

$$\begin{aligned} \omega_i &= \frac{d\theta}{dt} \\ d\theta &= \omega_i dt \\ \theta &= \int_0^t \omega_i dt = \int_0^t [\omega_c + K V_m \cos \omega_m t] dt \\ &= \omega_c \int_0^t dt + K V_m \int_0^t \cos \omega_m t dt = \omega_c t + \frac{K V_m}{\omega_m} \sin \omega_m t = \omega_c t + \frac{\delta_{max}}{\omega_m} \sin \omega_m t \quad \dots(6) \\ &= \omega_c t + M_F \sin \omega_m t \end{aligned}$$

put (6) in (5)

$$v_{fm}(t) = V_c \sin [\omega_c t + M_F \sin \omega_m t] \quad \dots(7)$$

where,  $M_F = \frac{\delta_{max}}{\omega_m}$  – Modulation Index

